



We1C-3



A 3-Way GaN Doherty Power Amplifier for 28GHz 5G FR2 Operation

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Outline



- Introduction
- Motivation
- Theory review
- Design
- Results
- Comparison with state of the art
- Conclusion
- Acknowledgements





Introduction



Wireless communication systems:

- High-PAPR signals (9-12 dB)
- FR1 band well-established
- Foreseen extension towards FR2 bands
- Power amplifiers:
 - Back-off efficiency enhancement
 - Gain requirement vs. achievable bandwidth
 - Single-stage vs. multi-stage implementation





Introduction



- Load-modulated PAs for deep back-off:
 - Doherty (asymmetric, multi-way)
 - DEPA
 - S-LMBA
- Challenges:
 - Gain penalty (DPA, S-LMBA)
 - Reliability (DEPA, S-LMBA, DPA?)
 - Number of active devices (DEPA)
 - Bandwidth

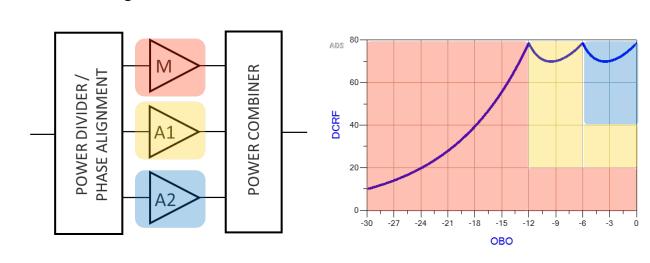




Motivation



- 2-way DPA: 6-dB back-off
 - Popular and established at < 6 GHz
 - Recently demonstrated at > 6 GHz (18 GHz, 28 GHz)
- A-way Doherty PA architecture: deep back-off
 - Combiner design for improved reliability
 - no Main overdrive
 - Deal with gain penalty
 - Multi-stage implementation
 - Extension towards FR2 bands



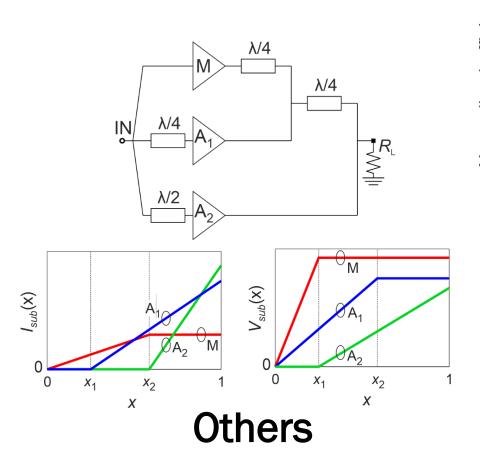


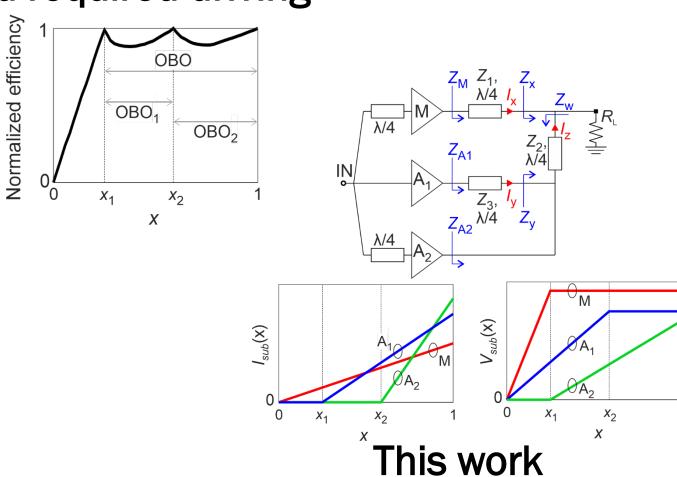


Theory review



Combiner topology and required driving









Theory review

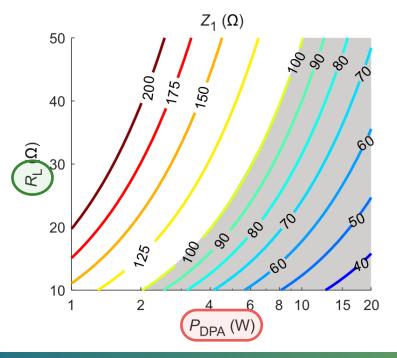


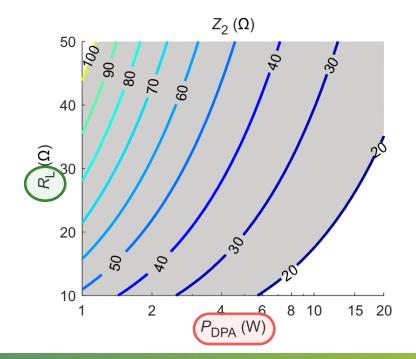
Design space

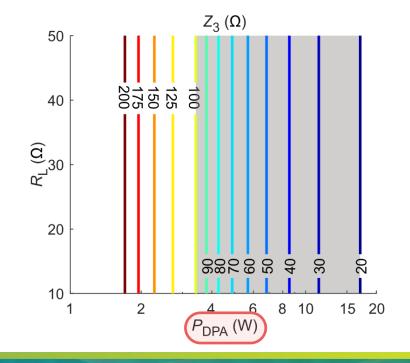
$$Z_{1} = \frac{V_{\text{Max,M}}}{I_{x}(1)} = \frac{V_{\text{Max,M}}}{\alpha_{1}} \cdot \sqrt{\frac{R_{L}}{2 P_{\text{DPA,sat}}}}$$

$$Z_{2} = \frac{V_{\text{Max,A}_{2}}}{I_{z}(1)} = \frac{\beta_{2} \cdot V_{\text{Max,M}}}{1 - \alpha_{1}} \cdot \sqrt{\frac{R_{L}}{2 P_{\text{DPA,sat}}}} \qquad Z_{3} = \frac{V_{\text{Max,A}_{1}}}{I_{y}(x_{2})} = \frac{\beta_{1} \cdot \beta_{2} \cdot V_{\text{Max,M}}^{2}}{2 \cdot \alpha_{2}(1 - \alpha_{1}) P_{\text{DPA,sat}}}$$

$$Z_3 = \frac{V_{\text{Max,A}_1}}{I_{\text{y}}(x_2)} = \frac{\beta_1 \cdot \beta_2 \cdot V_{\text{Max,M}}^2}{2 \cdot \alpha_2 (1 - \alpha_1) P_{\text{DPA,sat}}}$$





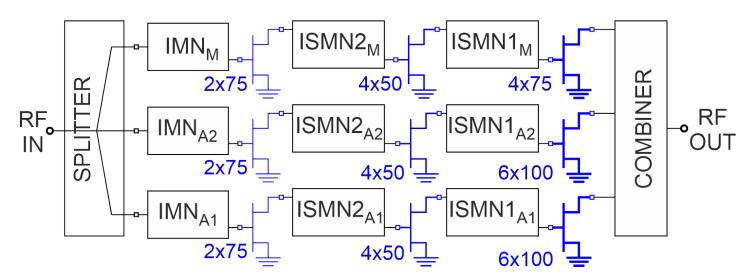








- Technology
 - WIN Semi's NP15 (150 nm GaN/SiC HEMT)
- Specifications
 - Sat. output power: 3 W
 - Linear gain: 20 dB
 - OBO: 6dB / 12dB



- Architecture
 - 3-stage (pre-driver + driver + final)

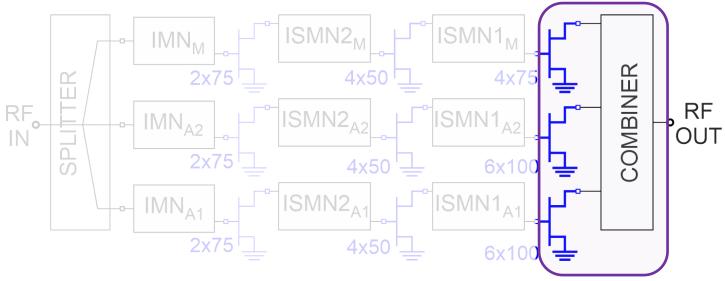


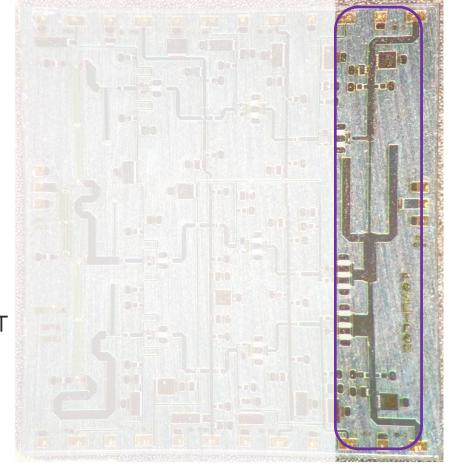




Combiner

- Semi-lumped $\lambda/4$
- Line-stub post-matching
 - from $\approx 11 \Omega$ to 50Ω





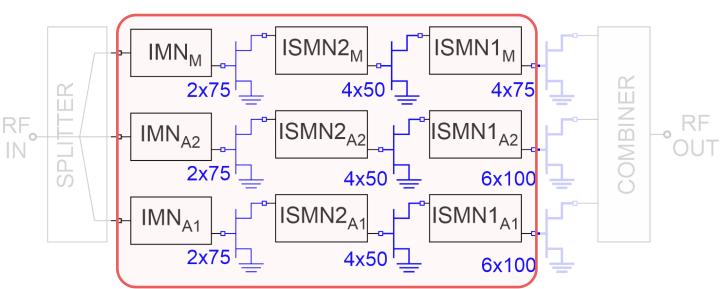


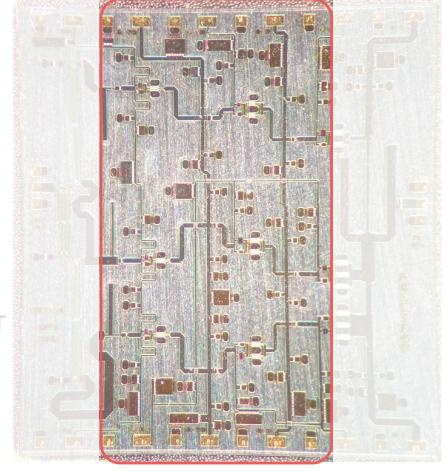




Symmetric MNs

- Simple routing
- Trade-off with degrees of freedom



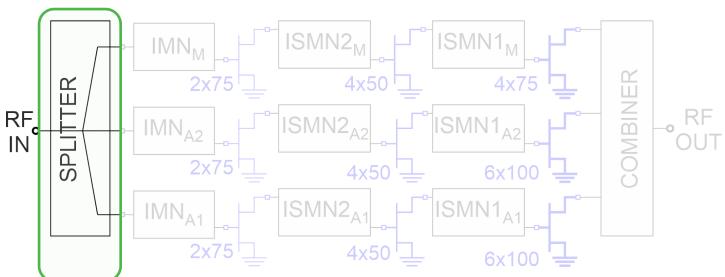


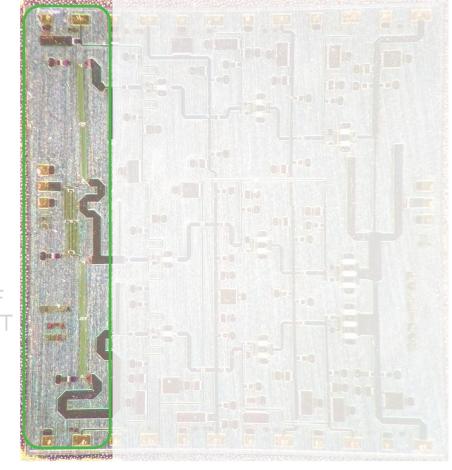






- Input splitter
 - 3-way
 - Isolated
 - 2 levels of Lange couplers







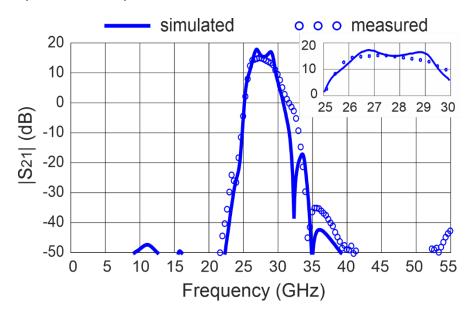


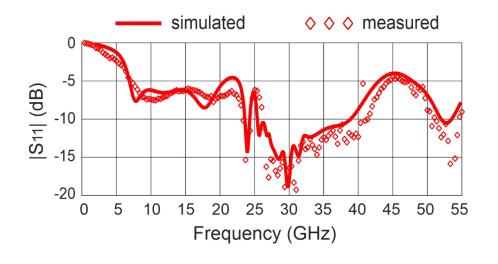
Results

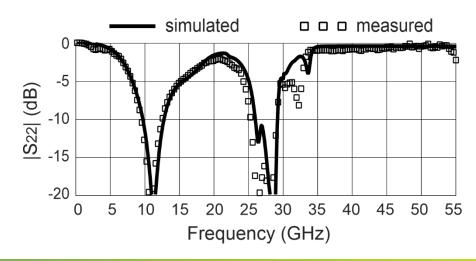


Small signal

$$V_{DD}=20 V$$
 $V_{G,pd}=V_{G,dM}=V_{G,M}=-2 V \rightarrow I_D \approx 30 \text{ mA}$
 $V_{G,dA1}=V_{G,A1}=-2.2 V \rightarrow I_D=0 \text{ mA}$
 $V_{G,dA1}=V_{G,A1}=-3.4 V \rightarrow I_D=0 \text{ mA}$









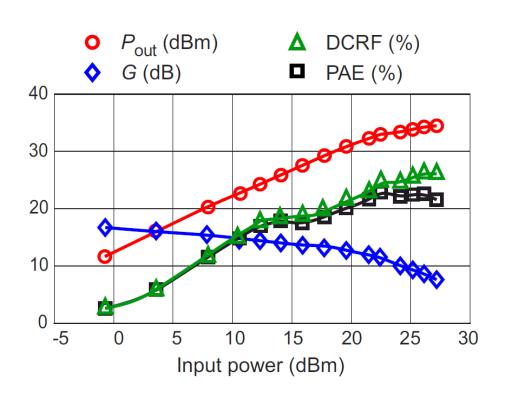


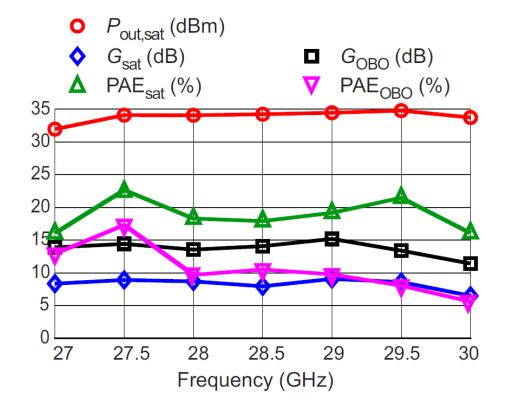
Results



Large signal: 1-tone CW

27.5 GHz







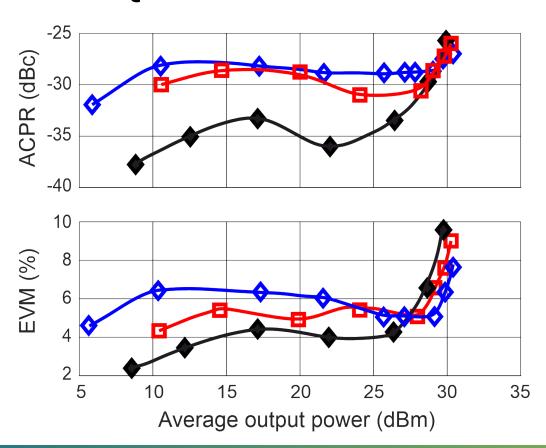


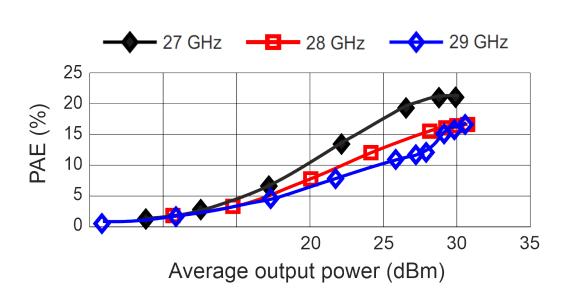
Results



Large signal: modulated

- 64-QAM standard NR downlink signal, 50 MHz BW, 10 dB PAPR









Comparison with SoA



Techn.	Freq. (GHz)	P _{out,sat} (dBm)	PAE _{sat} (%)	PAE [#] OBO (%)	G _{sat} (dB)	Ref
Si	28	22.4	40*	18*	10*	[10]
Si	26	18.7	32*	8*	12.5*	[11]
GaAs	27.25–29.75	27	37	12*	18	[9]
GaAs	28	30	38	10*	15*	[12]
GaN	26–30	36.1	26.7	15*	7	[13]
GaN	24–28	34	23	-	18	[14]
GaN	28.5–29.5	29*	30*	18*	4*	[15]
GaN	28–29	34	20	12	8	[8]
GaN	27.5–29.5	34	18–23	8–16	8–9	T. W.

[#] PAE at 10 dB OBO





^{*} Value extrapolated from graphs

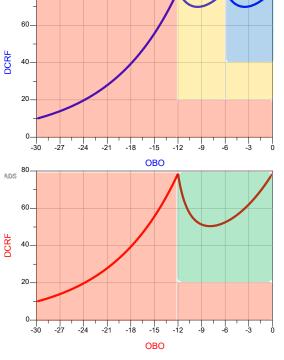


Conclusion



- Theory and design space
 - 3-way DPA for deep back-off efficiency enhancement
 - Extension to N>3 possible
 - No Main overdrive
- Implementation
 - MMIC 3-stage 3-way DPA
 - Challenge of FR2 frequencies: gain
 - Good efficiency, but can be improved
 - High intrinsic linearity









Acknowledgements



- Technology:
 - WIN Semiconductors: University Program
 - Italian Universities Consortium (MECSA)





- Instrumentation:
 - Keysight Technologies



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 - Italian Ministry of Universities and Research (MUR)



