

**WE3C-3**

# **A GaN Gain Enhancement PA with Peak Power Combining**

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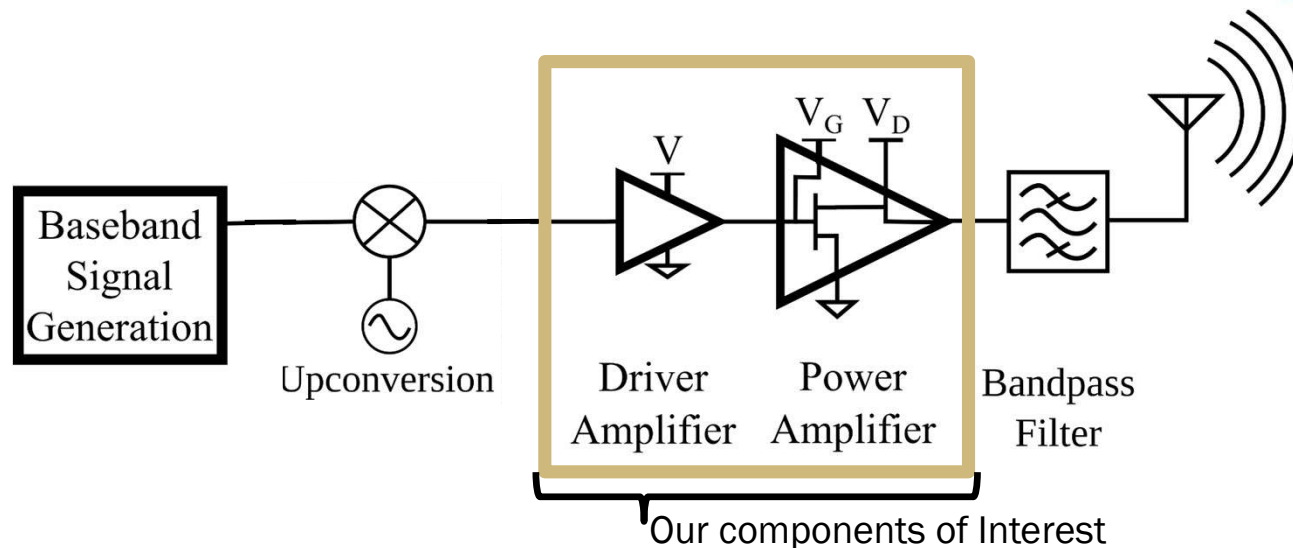
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# Outline of Discussion

1. Context, Motivation, and Important Design Metrics
2. Gain Enhancement PA (GEPA) Concept and Theory
3. Initial GEPA Prototype Design and Results
4. Conclusions

# RF Communications Transmitter System Context

There are many kinds of transmitter structure depending on the application. This represents a general transmitter, or single element in a phased array.

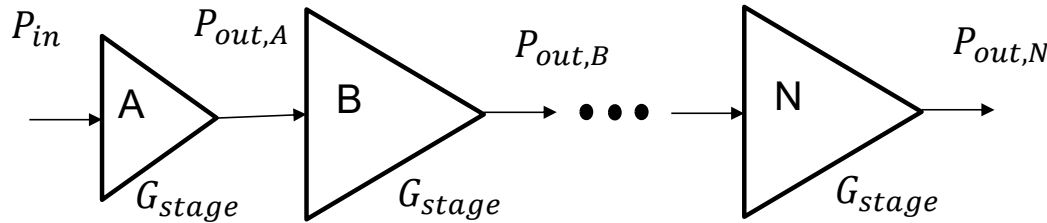


Canonically the performance of Power Amplifier (PA) determines the efficiency as measured by drain efficiency ( $\eta_D$ ) or power added efficiency (PAE) of a given transmitter.

HOWEVER, what happens when the PA has low gain and multiple drivers are needed, as in mmWave applications?

# IMS Amplifier Cascade Issues – Drain Efficiency

Assume all gain stages have the same peak drain efficiency, identical gain, and that they draw dc power proportional to output power.



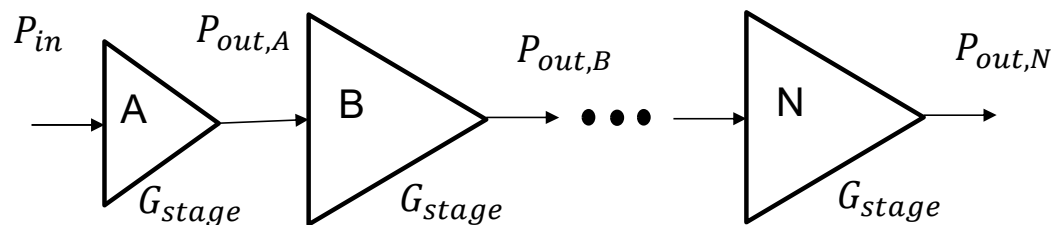
$$P_{dc,n} = P_{dc,n-1} / G_{stage}$$

$$\eta_{D,tot,2} = \frac{P_{out,B}}{P_{dc,B} + P_{dc,A}} = \frac{P_{out,B}}{P_{dc,B} \left(1 + 1/G_{stage}\right)} = \eta_{D,B} \frac{G_{stage}}{G_{stage} + 1}$$

$$\eta_{D,tot,3} = \frac{P_{out,C}}{P_{dc,C} + P_{dc,B} + P_{dc,A}} = \frac{P_{out,C}}{P_{dc,C} \left(1 + 1/G_{stage} + 1/G_{stage}^2\right)} = \eta_{D,C} \frac{G_{stage}^2}{G_{stage}^2 + G_{stage} + 1}$$

$$\eta_{D,tot,N} = \frac{P_{out,N}}{\sum_{n=0}^N P_{dc,n}} = \frac{P_{out,N}}{P_{dc,N} \sum_{n=0}^N 1/G_{stage}^n} = \eta_{D,N} \frac{G_{stage}^N}{\sum_{n=0}^N G_{stage}^n} = \eta_{D,N} \frac{(G_{stage} - 1)G_{stage}^N}{G_{stage}^{N+1} - 1}$$

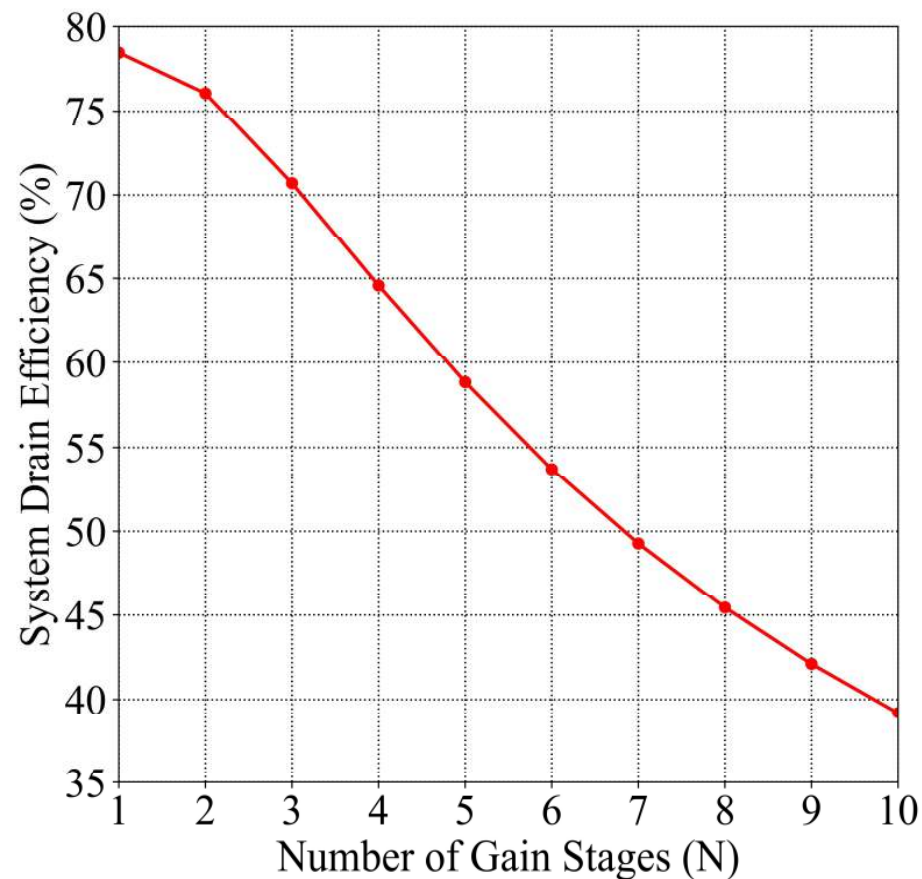
Assume all gain stages have the same peak drain efficiency, identical gain, and that they draw dc power proportional to output power.



$$\eta_{D,tot} = \eta_{D,N} \frac{(G_{stage} - 1)G_{stage}^N}{G_{stage}^{N+1} - 1}$$

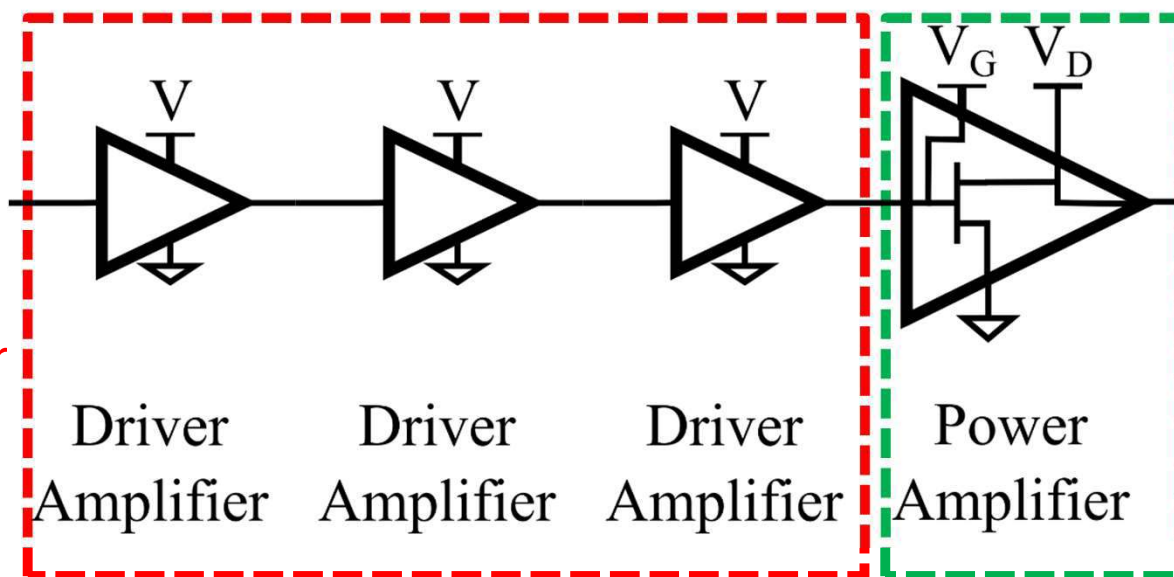
Adding amplifier stages with low gain reduces overall efficiency!  
 Conventional assumption that final stage is dominant is not true.

Total cascaded gain: 30 dB



# Amplifier Cascade Issues - Drivers

Must be designed with drive margin to ensure the PA reaches peak output power – guaranteed lower efficiency!



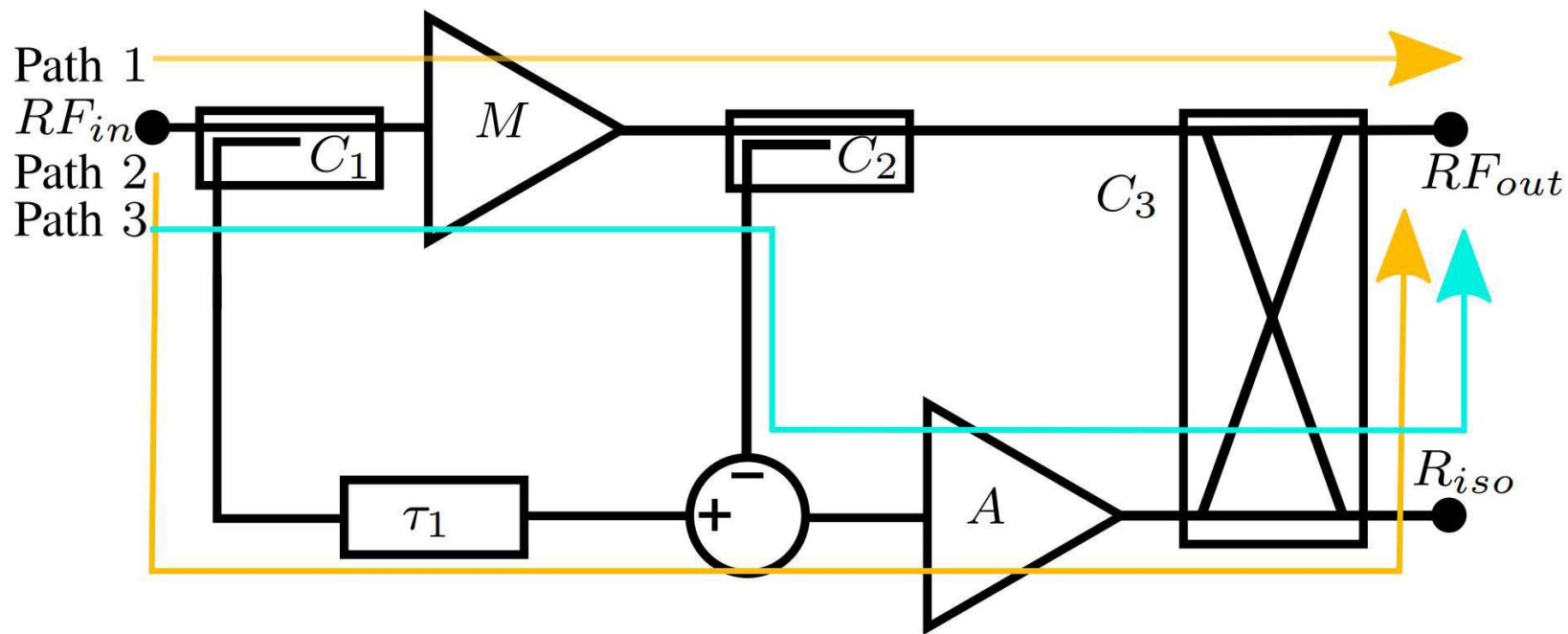
Optimized for efficiency at peak output power

Each additional driver amplifier stage adds:

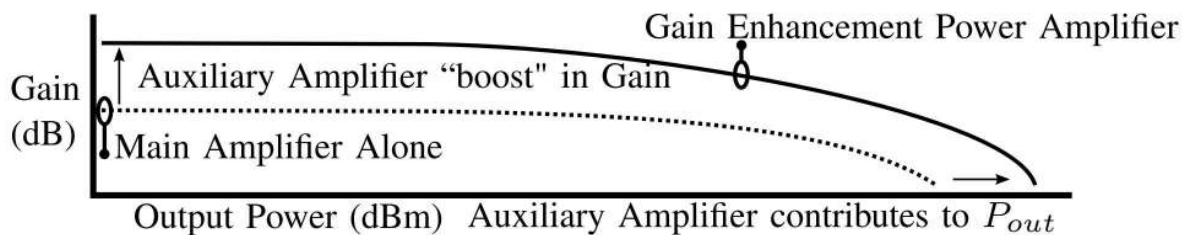
- Increases Gain
- Increases Current Draw (decreasing efficiency)
- Adds Additional Failure Points

How can we minimize the number of driver amplifiers without harming PA efficiency?

# Gain Enhancement PA Concept



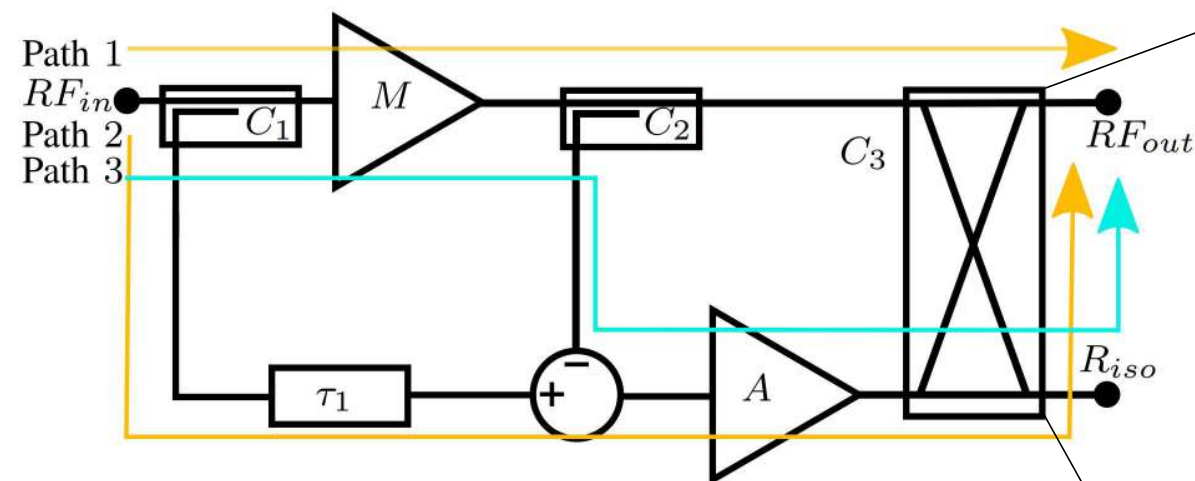
Series connection increases gain of GEPA structure!



Parallel connection ensures output of both amplifiers contributes to overall output!



# Design Assumptions and Coupler $C_3$



Coupler  $C_3$  is implemented as a 3dB quadrature hybrid.

$C_3$  has a coupling factor of 3dB as the main and auxiliary amplifiers are identical output power.

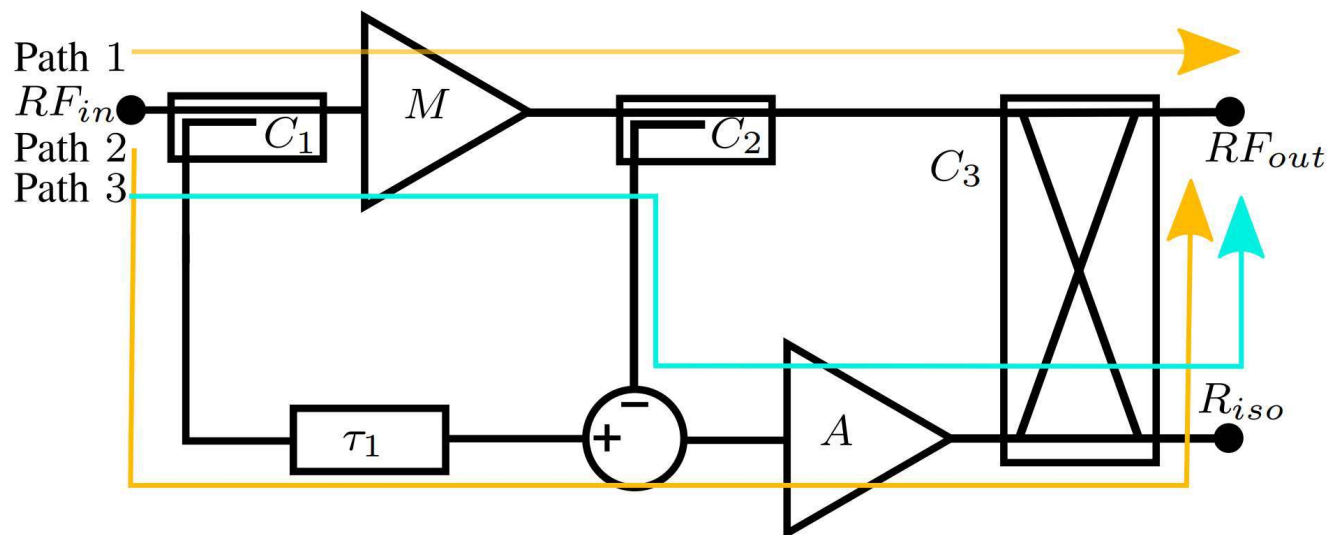
$C_3$  is a quadrature coupler to take advantage of that structure's natural resilience to impedance variation, a useful property in phased array applications.

## Core Design Assumptions:

- The Main and Auxiliary amplifiers are identical ( $G = G_M = G_A$ ), ( $P_{out,max} = P_{out,max,M} = P_{out,max,A}$ )
- Subtractor element is ideal (non-idealities absorbed into surrounding components)



# Phase Alignment

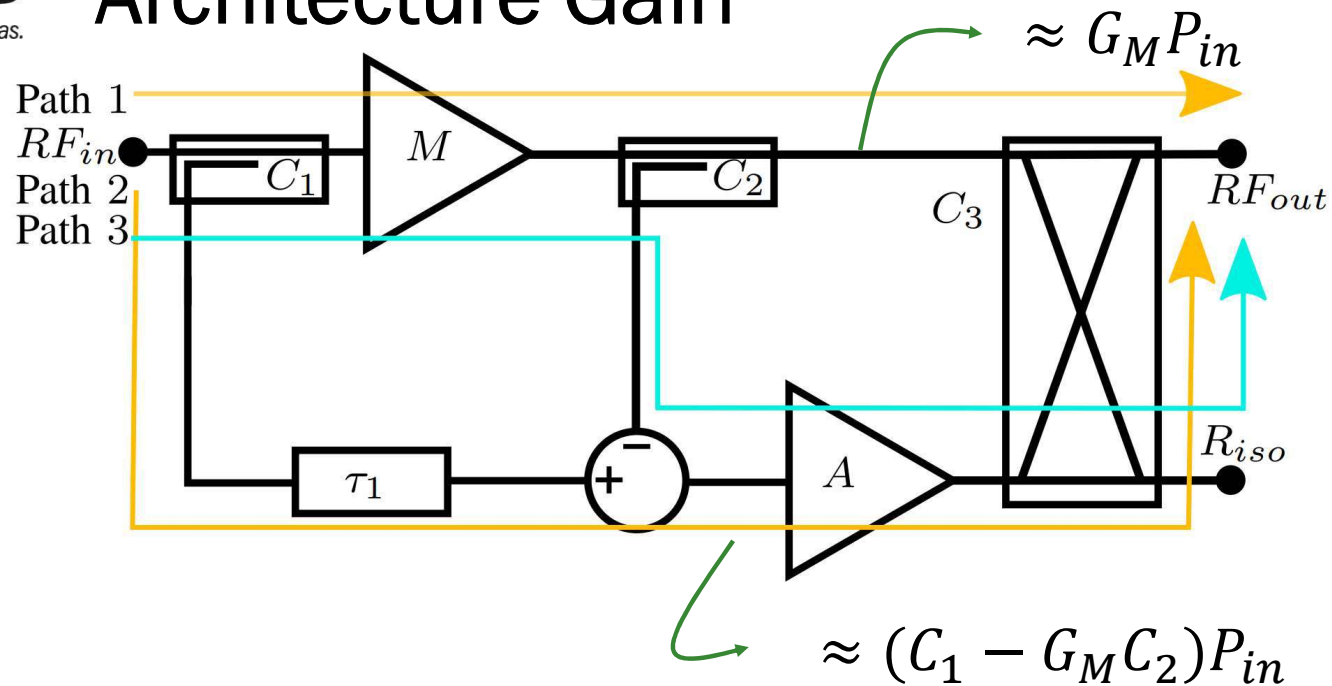


Path 1 and 2 must have equal phase to add at the output  
 Path 3 must differ from Path 2 by  $180^\circ$

$$\theta_{P1} = \theta_{P2}$$

$$\theta_{P2} = \theta_{P3} - \pi$$

## Architecture Gain

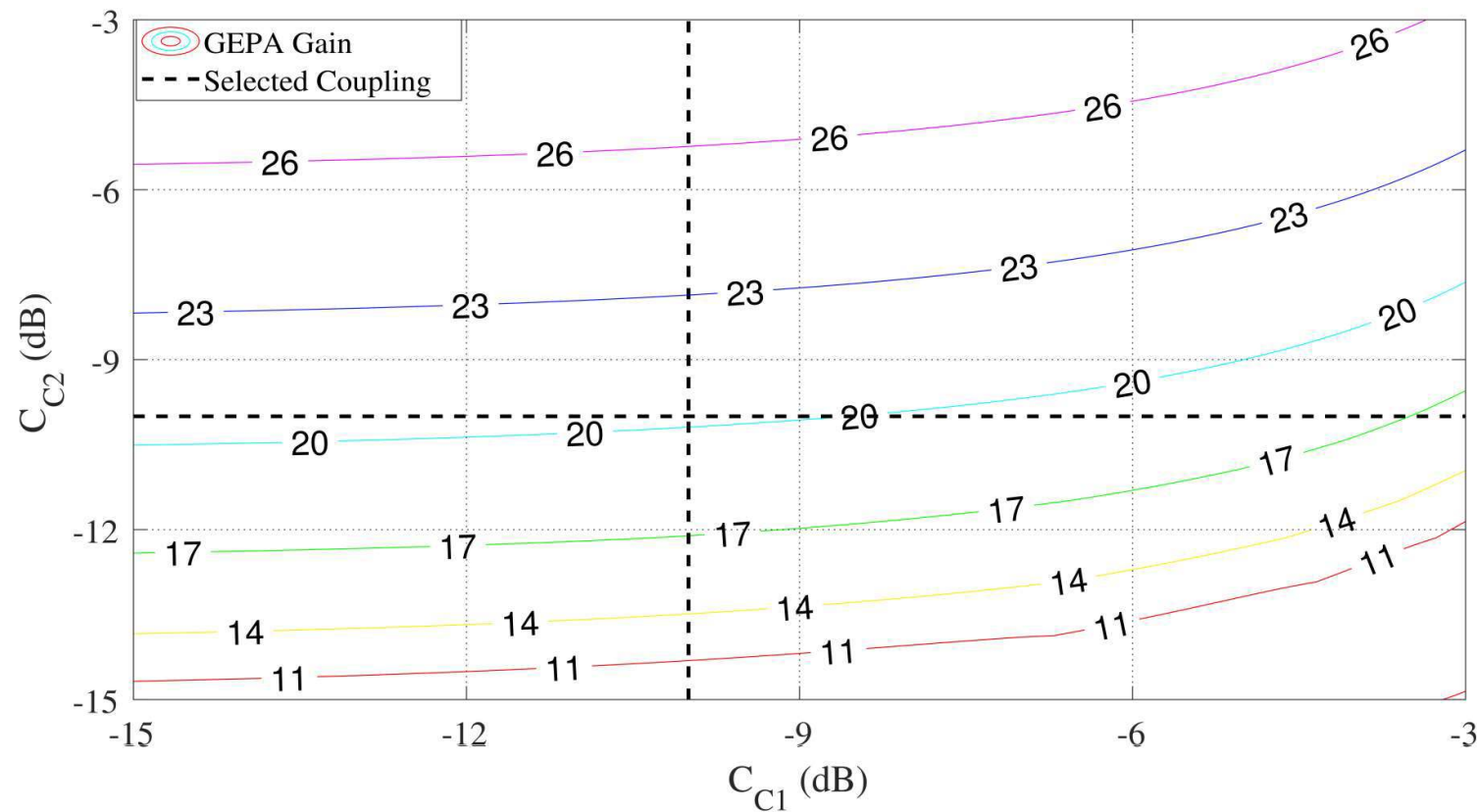


Full gain expression:

$$G_{GEPA} = |G(1 - C_{C2} + C_{C1}C_{C2}) + G^2(C_{C1}C_{C2} - C_{C2})|$$

# Selection of Couplers $C_1$ and $C_2$

Branch Amplifier Gain of 16 dB:



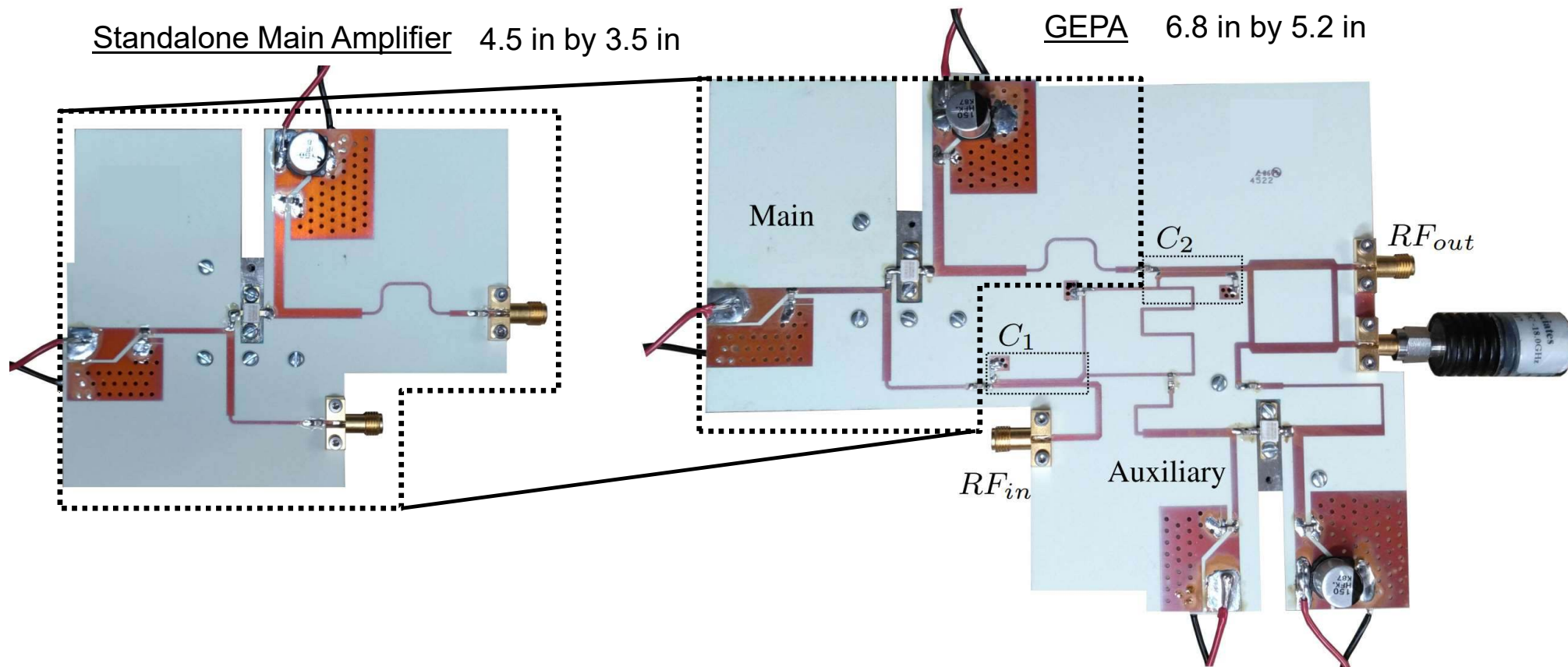
10 dB coupling factors selected in this work such that  $C_{C1} = C_{C2}$  and to ensure realizability in RO4350B.

$$G_{GEPA} = |G(1 - C_{C2} + C_{C1}C_{C2}) + G^2(C_{C1}C_{C2} - C_{C2})|$$

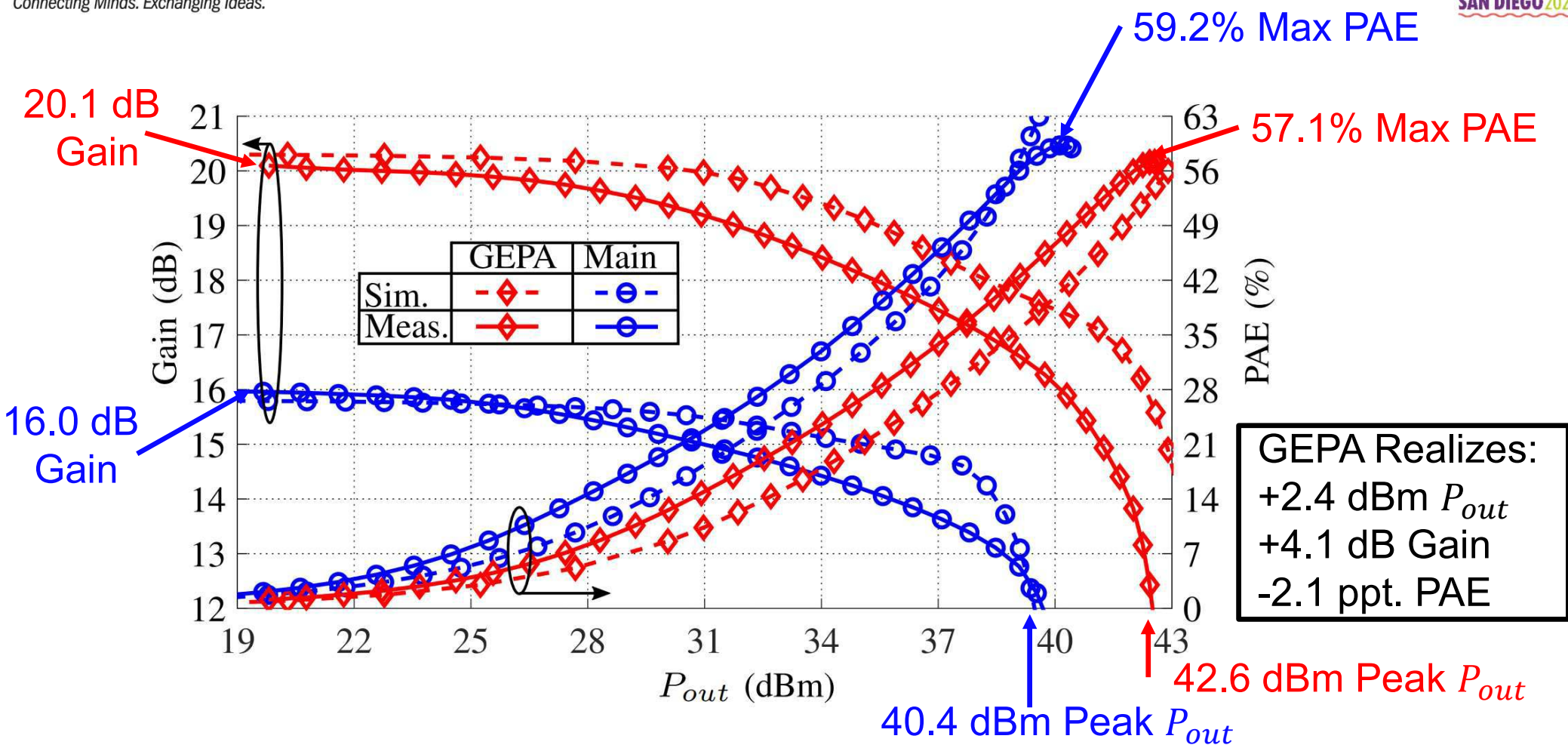
# Final Prototype at 2.03 GHz

Standalone Main Amplifier 4.5 in by 3.5 in

GEPA 6.8 in by 5.2 in

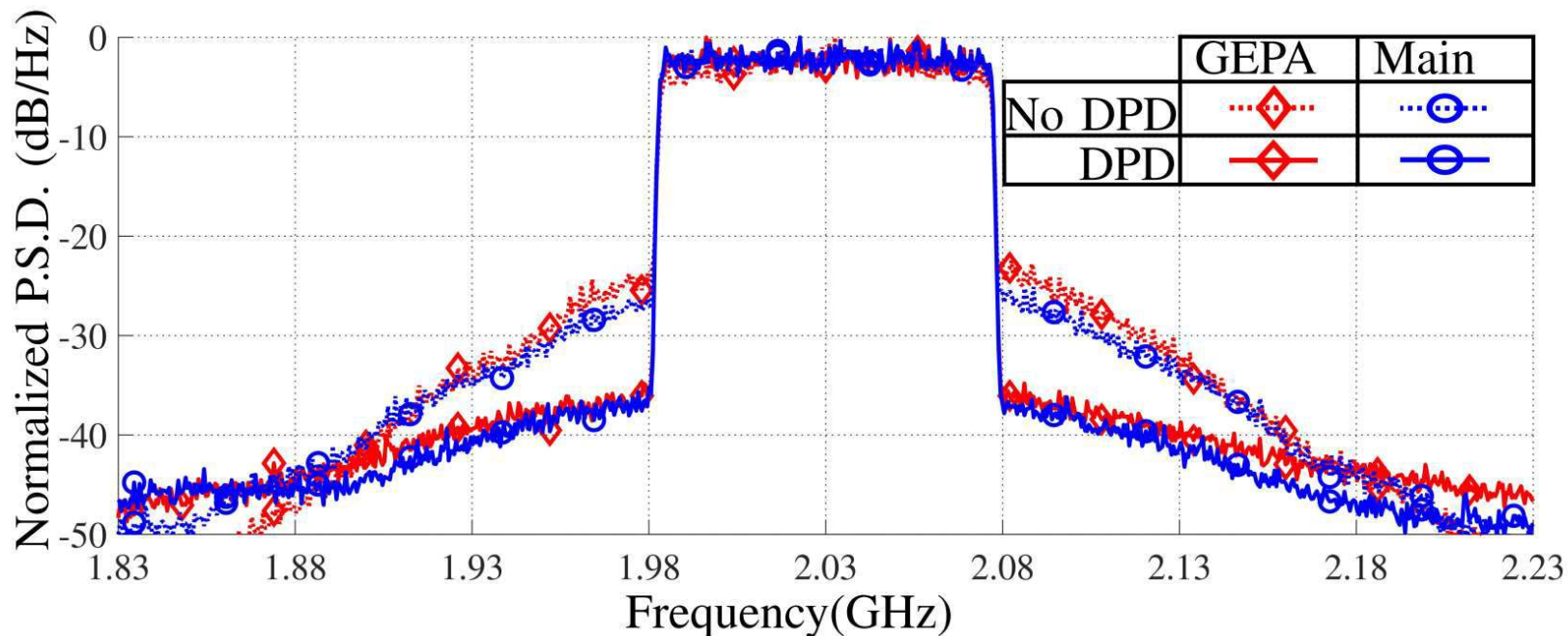


# CW Measured Results at 2.03 GHz





# LTE Measured Results at 2.03 GHz



Measured output spectrum for a 100 MHz wide LTE-like signal with 10 dB PAPR

	Peak Output Power (dBm)	Average Drain Efficiency (%)	Average Input Power (dBm)
GEPA (w/DPD)	42.2	26.7	14.5
Standalone Main (w/DPD)	39.9	29.2	18.5

# Comparison to State-of-the-Art

Ref.	Year	Arch.	Freq. (GHz)	Gain (dB)	$P_{out}$ (dBm)	Peak PAE (%)
W. Hallberg, M. Özen, and C. Fager, "Current scaled Doherty amplifier for high efficiency and high linearity," in IEEE MTT-S International Microwave Symposium, 2016, pp. 1–4.	2016	Doherty	2.14	11*	43.0	50*
P. H. Pednekar, W. Hallberg, C. Fager, and T. W. Barton, "Analysis and design of a Doherty-like RF-input load modulated balanced amplifier," IEEE Transactions on Microwave Theory and Techniques, vol. 66, no. 12, pp. 5322–5335, 2018.	2018	LMBA	2.4	12*	45.6	60.0
C. Shen, S. He, and X. Zhu, "Design of broadband high-efficiency Doherty power amplifier using post-matching network," in Asia-Pacific Microwave Conference, 2018, pp. 464–466.	2018	Doherty	2.1	12*	44*	57*
M. Jung, S. Min, and B.-W. Min, "A quasi-balanced power amplifier with feedforward linearization," IEEE Microwave and Wireless Components Letters, vol. 32, no. 4, pp. 312–315, 2022	2022	Balanced-FFA	3.5	16.5*	30.8	27.2
W. Sear, D. T. Donahue, M. Pirrone, and T. W. Barton, "Bias and bias line effects on wideband RF power amplifier performance," in IEEE Wireless and Microwave Technology Conference, 2022.	2022	Class-AB	2.2	14*	40.0	50*
<b>This Work</b>		Main Amplifier	2.03	16.0	40.4	59.2
		GEPA	2.03	20.1	42.6	57.1

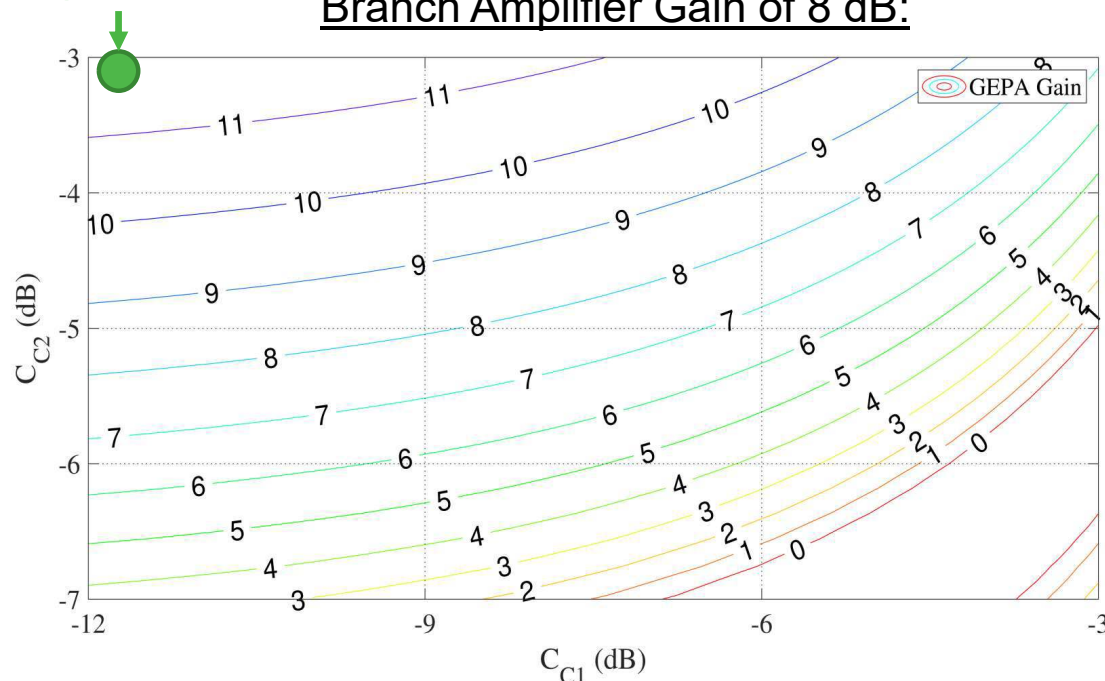
\* - read from graph



# Application to mmWave PAs

Optimal GEPA

Branch Amplifier Gain of 8 dB:



Consider the following Amplifier:

A. Der, W. Sear, Z. Popovic, G. Lasser, and T. Barton, "A S-C- / K-band reconfigurable GaAs MMIC power amplifier for 5G applications," in IEEE MTT-S International Microwave Symposium, 2021, pp. 873–876.

Operation at 22 GHz:

	Single Amplifier	2-Stage Cascade	Optimal GEPA*	GEPA-Cascade $\Delta$
$P_{out}$ (dBm)	20.7	20.7	23.7	+3
Gain (dB)	8	16	11.9	-3.1
PAE (%)	30	20.5	30	+9.5

\*The optimal GEPA targets maximal gain and assumes no combining loss.

mmWave PAs feature characteristically low per-stage gain, the perfect target for this technique!

# Summary of GEPA Performance

- ✓ ☒ Compound RF PA's connected simultaneously in parallel and in series (GEPA) can realize increased gain and output power compared to a single RF PA.
- ✓ ☒ This connection adds driver (main) amplifier power directly to the output at saturation,
- ✓ ☒ If both main and aux. amplifiers have identical RF performance the overall efficiency of the structure will be that efficiency, minus combining losses.
- ✓ ☒ The structure is linearizable with DPD.

Thank you all for your time and attention!  
Questions and comments are appreciated!



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